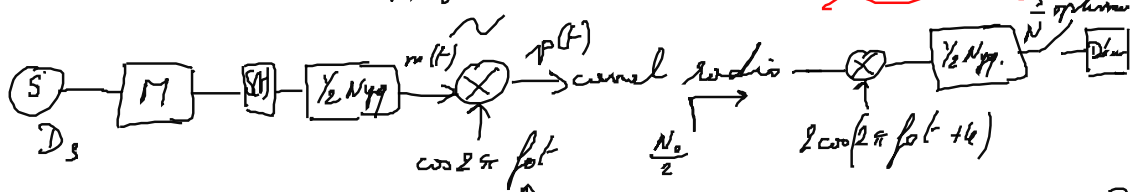


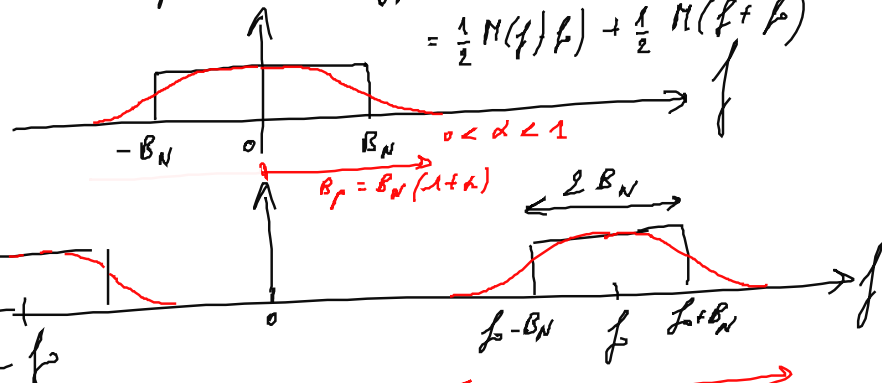
leo modulations



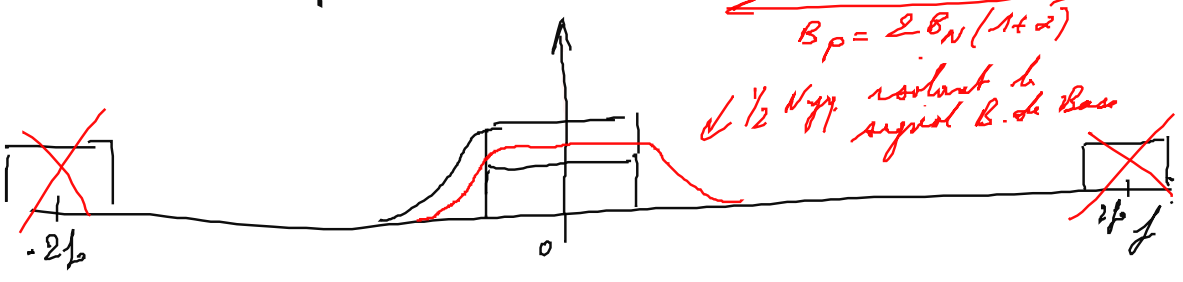
demodulatore

$$m(t) \cdot \cos(2\pi f_0 t) \rightarrow P(f) = M(f) * \frac{1}{2} [\delta(f-f_0) + \delta(f+f_0)]$$

$$= \frac{1}{2} M(f-f_0) + \frac{1}{2} M(f+f_0)$$



$B_p = 2 B_N (1 + \alpha)$
 \downarrow 1/2 Nyq isolant h. signal B. de Base



Effet du déphasage φ de l'oscillateur local (L.O.)

on reçoit $p(t) = m(t) \cos 2\pi f_c t$

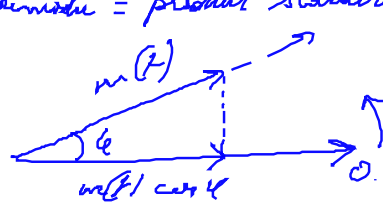
O.L.

on demodule $\rightarrow p(t) \cdot 2 \cos(2\pi f_c t + \varphi)$

$$= m(t) \cdot \underbrace{\cos 2\pi f_c t}_a \cdot \underbrace{2 \cos(2\pi f_c t + \varphi)}_b$$

$$\cos(a \pm b) = \cos a \cos b \mp \sin a \sin b$$

demodu = produit scalaire = projections



$$= m(t) \left[\cos(4\pi f_c t + \varphi) + \cos \varphi \right]$$

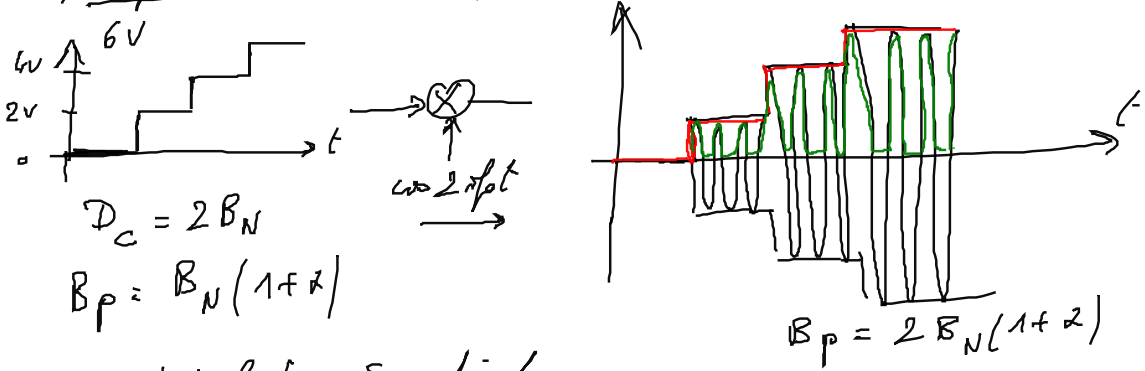
$$= \underbrace{m(t) \cos(2\pi \cdot 2f_c t + \varphi)}_{\text{terme centré sur } 2f} + \underbrace{m(t) \cos \varphi}_{\text{terme à la base}}$$

Après le $\frac{1}{2}$ Nyquist

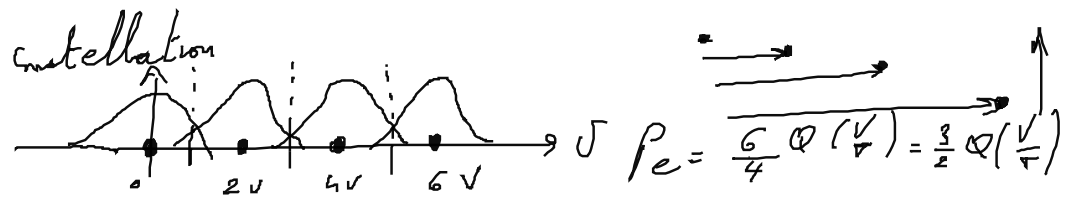
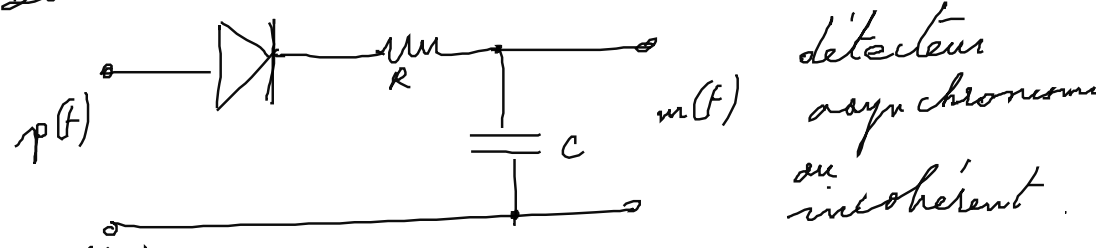
il reste $m(t) \cos \varphi < m(t)$ $0 < \cos \varphi < 1$
 Si mauvais asservissement de phase du récepteur, on a une perte de niveau sur $m(t)$ et donc une dégradation du $\frac{S}{N}$ avant échantillonnage

Les différents modulations :

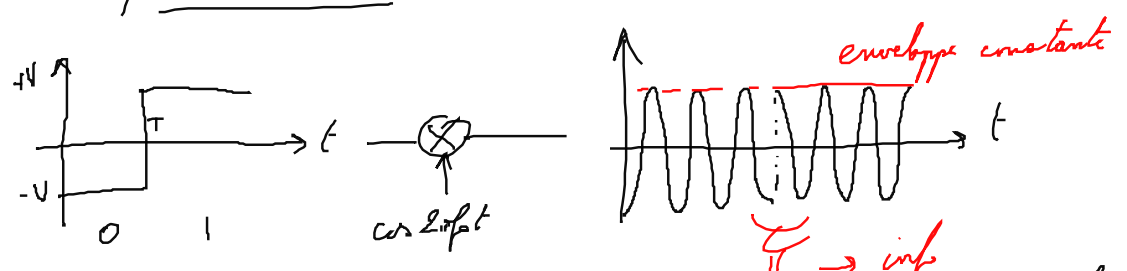
a) Amplitude (ASK) Amplitude Shift Keying



Démodulateur à diode :

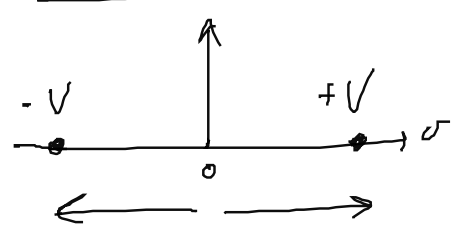


b) La BPSK (Modulation de phase binaire)



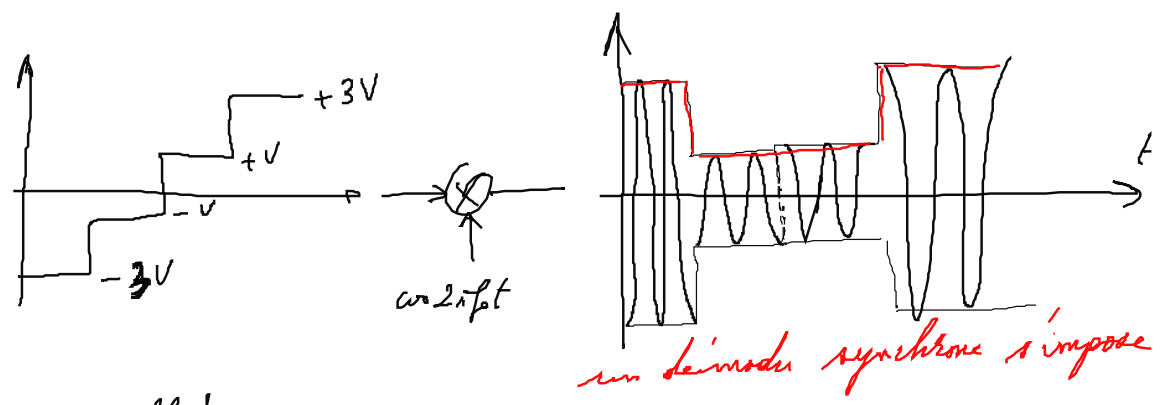
nécessite obligatoirement d'un détecteur synchrone
(multiplicateur + O. Local asservi en phase ou cohérent)

constellation :

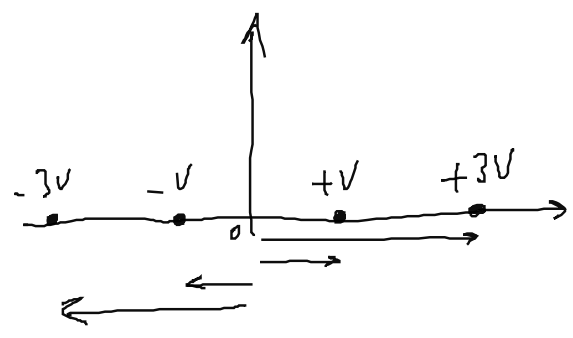


$$P_e = Q\left(\frac{V}{\sigma}\right) = Q\left(\sqrt{\frac{2E_B}{N_0}}\right)$$

c) CS-APK (Amplitude and Phase Shift Keying)

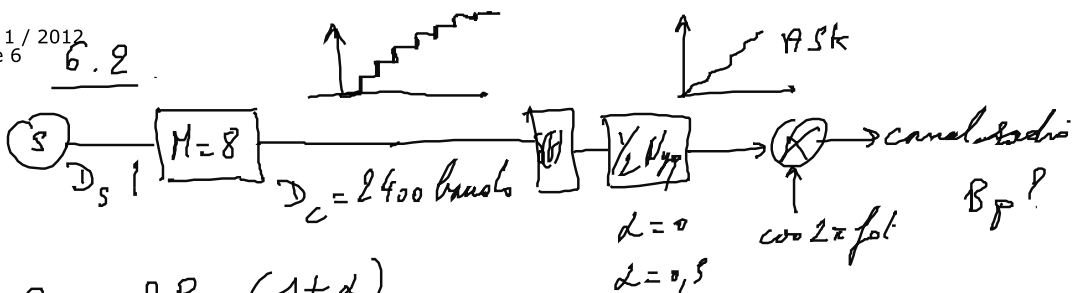


Constellation



$$P_e = \frac{6}{4} Q\left(\frac{V}{T}\right)$$

$$= \frac{3}{2} Q\left(\frac{V}{T}\right)$$



$$B_p = 2B_N (1 + \alpha)$$

$$D_c = 2400 \text{ bauds} = 2B_N \text{ (critère de Nyquist)}$$

$$2B_N = 2400 \text{ Hz}$$

a) $\alpha = 0$

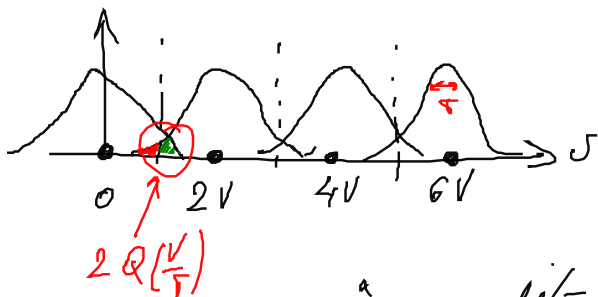
$$B_p = 2B_N = 2400 \text{ Hz}$$

b) $\alpha = 0,5$

$$B_p = 2400 (1 + 0,5) = 3600 \text{ Hz}$$

Calcul du taux d'erreur pour la 4-ASK

constellation:



$$\begin{aligned}
 P_e &= \frac{6}{4} Q\left(\frac{V}{T}\right) \\
 &= \frac{3}{2} Q\left(\frac{V}{T}\right) \\
 &= \frac{2(M-1)}{M} Q\left(\frac{V}{T}\right)
 \end{aligned}$$

grâce au filtrage adapté:

$$P_e = \frac{3}{2} Q\left(\sqrt{\frac{2V^2 T_{eq}}{N_0}}\right)$$

Retour au signal:

$$P_{\text{moy}} = \frac{(0)^2 + (2V)^2 + (4V)^2 + (6V)^2}{4} \rightarrow \sum x^2 p(x)$$

$$= \frac{4V^2 + 16V^2 + 36V^2}{4} = \frac{56V^2}{4} = 14V^2$$

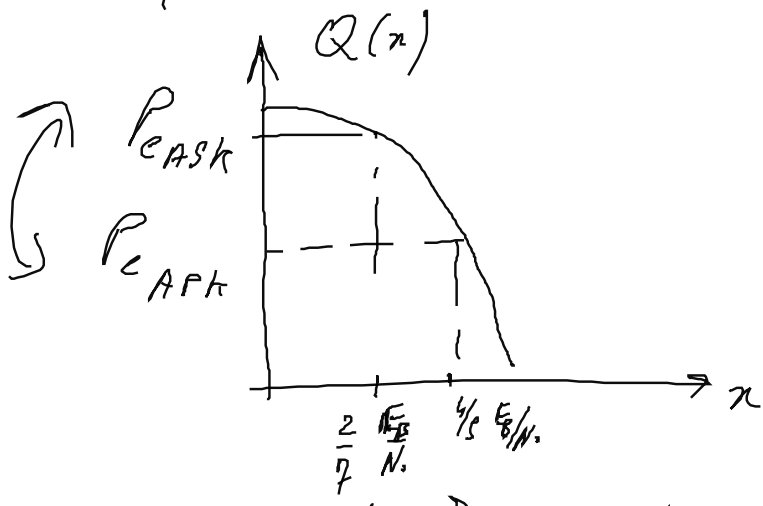
$$\bar{E}_s = P_{\text{moy}} \cdot T_{eq} = 14V^2 T_{eq} \quad E_B = \frac{\bar{E}_s}{\log_2 M} = \frac{14V^2}{2} = 7V^2 \quad \text{J/bit}$$

$$V^2 T_{eq} = \frac{1}{7} E_B \Rightarrow P_{e_{\text{bit}}} = \frac{3}{2} Q\left(\sqrt{\frac{2}{7} \frac{E_B}{N_0}}\right) \rightarrow P_{e_{\text{bit}}} = \frac{3}{2} Q\left(\sqrt{\frac{2}{7} \frac{E_B}{N_0}}\right)$$

Comparison 4-ASK / 2-APK

$$P_{e\text{bit}} = \frac{3}{4} Q\left(\sqrt{\frac{2}{7} \frac{E_B}{N_0}}\right)$$

$$P_{e\text{bit}} = \frac{3}{4} Q\left(\sqrt{\frac{4}{5} \frac{E_B}{N_0}}\right)$$



À P_e égale, il faut augmenter E_B de l'ASK pour rattrapper l'APK → moins bon bilan énergétique pour l'ASK

6.9

$$D_s = 64 \text{ kbit/s} \rightarrow 8\text{-ASK} \quad \alpha = 0,33$$

B_F !

$$B_p = 2 B_N (1 + \alpha)$$
$$= D_c (1 + \alpha) \quad (\text{Nyquist})$$

$$= \frac{D_s}{\log_2 M} (1 + \alpha)$$

$$= \frac{64000}{3} (1 + 0,33) = 28373,33 \text{ Hz}$$
$$\approx 28 \text{ kHz}$$

7.1 BPSK $B_p = 8000 \text{ Hz}$
 $\alpha = 0$
 $D_s \text{ max ?}$

$$B_p = 2 B_N \cdot (1 + \alpha)$$

$$8000 = 2 B_N (1 + 0)$$
$$= 2 B_N$$

$$= D_c \text{ (Nyquist)}$$

$$D_c = 8000 \text{ bauds}$$

$$D_s = D_c \cdot \log_2 M \quad M = 2 \text{ (Binary PSK)}$$
$$= 8000 \text{ bits/s}$$

7.4. $D_S = 2,08 \text{ Mbits/s}$ $P_{e \text{ bit}} \leq 10^{-6}$

$\frac{N_0}{2} = 10^{-8}$ $\alpha = 1$

$B_p?$ $S?$

a) Bande de base ($M=2$)

$D_S = D_c \log_2 M \rightarrow D_S = D_c \rightarrow D_c = 2,08 \text{ Mbauds}$

$D_c = 2B_N \rightarrow 2B_N = 2,08 \text{ MHz}$

$B_p = 2B_N(1+\alpha) = 2,08 \cdot 10^6 (1+1)$
 $= 4,16 \text{ MHz}$

$P_{e \text{ bit}} = Q\left(\sqrt{\frac{2E_B}{N_0}}\right) < 10^{-6} \Rightarrow \sqrt{\frac{2E_B}{N_0}} = x = 4,75$

$\frac{E_B}{N_0} = 4,75^2 \cdot \frac{1}{2} = 11,28$
 $\rightarrow 10,5 \text{ dB}$

$$E_B = \frac{E_B}{N_0} \cdot A_0 \quad \frac{N_0}{2} = 10^{-8}$$

$$= 11,28 \cdot 2 \cdot 10^{-8} = 22,56 \cdot 10^{-8}$$

$$S = E_B \cdot D_S = 22,56 \cdot 10^{-8} \cdot 2,08 \cdot 10^6 = 46,92 \cdot 10^{-2}$$

$$f_{bit} \times \text{bit/s} = 1/A = W \quad \approx 0,5 W$$

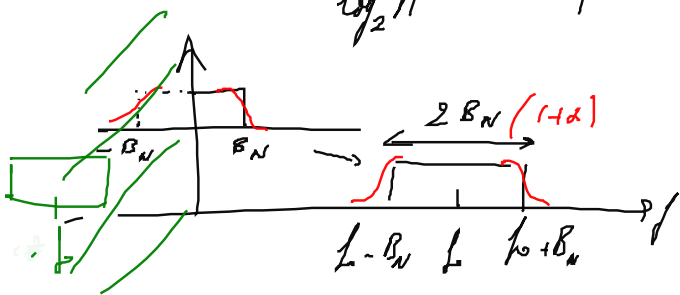
b) Caso de la 16 APK

$$B_P = 2B_N (1+\alpha) = 4B_N$$

$$D_C = \frac{D_S}{\log_2 M} = \frac{2,08 \cdot 10^6}{4}$$

$$B_P = \frac{2,08 \cdot 10^6}{4} \cdot 2$$

$$= 1,04 \text{ MHz}$$



Si nécessaire pour la 16-APK

$$P_{e_{\text{sympt}}} = \frac{2(M-1)}{M} Q\left(\sqrt{\frac{6}{M^2-1} \frac{E_B}{N_0} \cdot \log_2 M}\right)$$

$$= \frac{30}{16} Q\left(\sqrt{\frac{24}{255} \frac{E_B}{N_0}}\right)$$

$$= \frac{15}{8} Q\left(\sqrt{\frac{8}{85} \frac{E_B}{N_0}}\right) \leq 10^{-6}$$

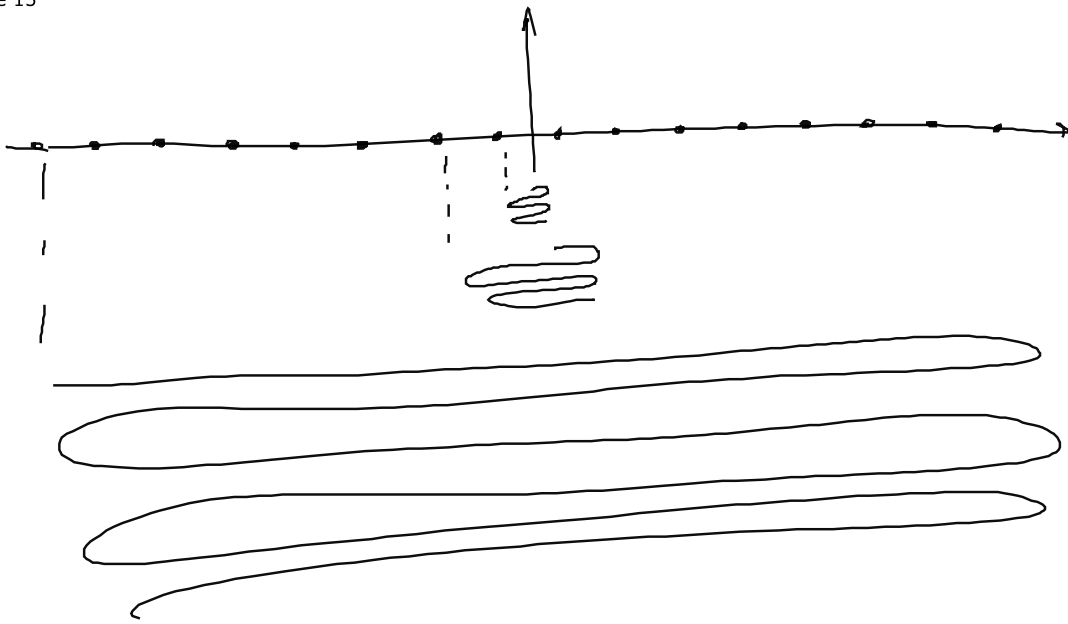
$$P_{e_{\text{bit}}} = \frac{P_{e_{\text{sympt}}}}{\log_2 M} = \frac{15}{32} Q\left(\sqrt{\frac{8}{85} \frac{E_B}{N_0}}\right) \leq 10^{-6}$$

$$Q\left(\sqrt{\frac{8}{85} \frac{E_B}{N_0}}\right) \leq 10^{-6} \cdot \frac{32}{15} = 2,13 \cdot 10^{-6}$$

$$\sqrt{\frac{8}{85} \frac{E_B}{N_0}} = 4,60 \Rightarrow \frac{E_B}{N_0} = 4,60^2 \cdot \frac{85}{8} = 224,8$$

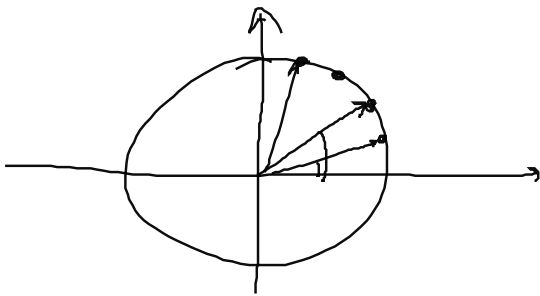
$$E_B = 224,8 \cdot 2 \cdot 10^{-8} = 449,6 \cdot 10^{-8} \text{ J/bit}$$

$$S = 449,6 \cdot 10^{-8} \cdot 2,08 \cdot 10^{+6} = 9,35 \text{ W}$$



c) Cas de la 16-PSK

M grande puissance
que pour APSK



$$P_e = 2 Q \left(\sqrt{2 \frac{E_B}{N_0} \cdot \frac{1}{2} M \cdot \sin \frac{\pi}{M}} \right)$$

$$P_{\text{chit}} = \frac{1}{2} Q \left(\sqrt{2 \frac{E_B}{N_0} \cdot 4 \cdot \sin \frac{\pi}{16}} \right) = \frac{1}{2} Q \left(\sqrt{8 \frac{E_B}{N_0} \sin \frac{\pi}{16}} \right)$$

$$Q \left(\sqrt{8 \frac{E_B}{N_0} \sin \frac{\pi}{16}} \right) \leq 2 \cdot 10^{-6} \leq 10^{-6}$$

$$\sqrt{8 \frac{E_B}{N_0} \sin \frac{\pi}{16}} = 4,61 \rightarrow \frac{E_B}{N_0} = \left(\frac{4,61}{\sin \frac{\pi}{16}} \right)^2 \cdot \frac{1}{8} = 69,8$$

$$E_B = 69,8 \cdot 2 \cdot 10^{-8} = 139,6 \cdot 10^{-8} \rightarrow S = \frac{139,6 \cdot 10^{-8} \cdot 2 \cdot 8 \cdot 10^6}{2,9 \text{ W}}$$